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Direct control of high magnetic fields for cold atom experiments based on NV centers

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Abstract

In ultracold quantum gases, the interactions between the individual atoms can be controlled by applying magnetic bias fields. As magnetic field fluctuations limit the precision here, typically a feedback loop needs to be employed to regulate the current through a pair of Helmholtz coils. No commercially available magnetic field sensor allows to measure large fields directly with high enough precision, leading to many unsatisfactory solutions being used in experiments. Here, we demonstrate a direct magnetic field stabilization in a regime previously not accessible, using NV centers as the magnetic field sensor. This allows us to measure and stabilize fields of 4.66 mT down to 12 nT RMS noise over the course of 24 h, measured on a 1 Hz bandwidth. We achieve a control of better than 1 ppm after 20 min of integration time, ensuring high long-term stability for experiments. This approach extends direct magnetic field control to strong magnetic fields, which could enable new precise quantum simulations in this regime.

1. Introduction

The direct control of modest magnetic fields up to 1 mT enabled experiments with cold atoms on diverse topics, like spin squeezing close to a Feshbach resonance [1,2], the quantum simulation of scaling dynamics far from equilibrium [3] or of a scalable building block for lattice gauge theories in atomic mixtures [4]. In all of these cases, the direct magnetic field control is based on active feedback from a fluxgate sensor, which limits this class of experiments to magnetic fields below 1 mT, the working range of fluxgate sensors [5]. Many phenomena, like droplet formation close to a heteronuclear Feshbach resonance [6] or spin changing collisions between fermionic and bosonic species [7], do however occur at much higher field strengths, reaching up to a few tens of mT. Due to a lack of suitable sensors in this regime, only the stabilization of the current that generates the magnetic field is feasible. However, this leaves out any magnetic field noise from external sources, like experiments in nearby labs, and power line 50 Hz noise. Thus, drifts of the magnetic field due to external sources hinder the long term stability, and prevent the acquisition of data over the course of several days, which is often necessary in quantum simulation using ultracold gases [8,9]. This shows the need for alternative sensors capable of measuring large magnetic fields with high precision, enabling precision experiments in this regime.

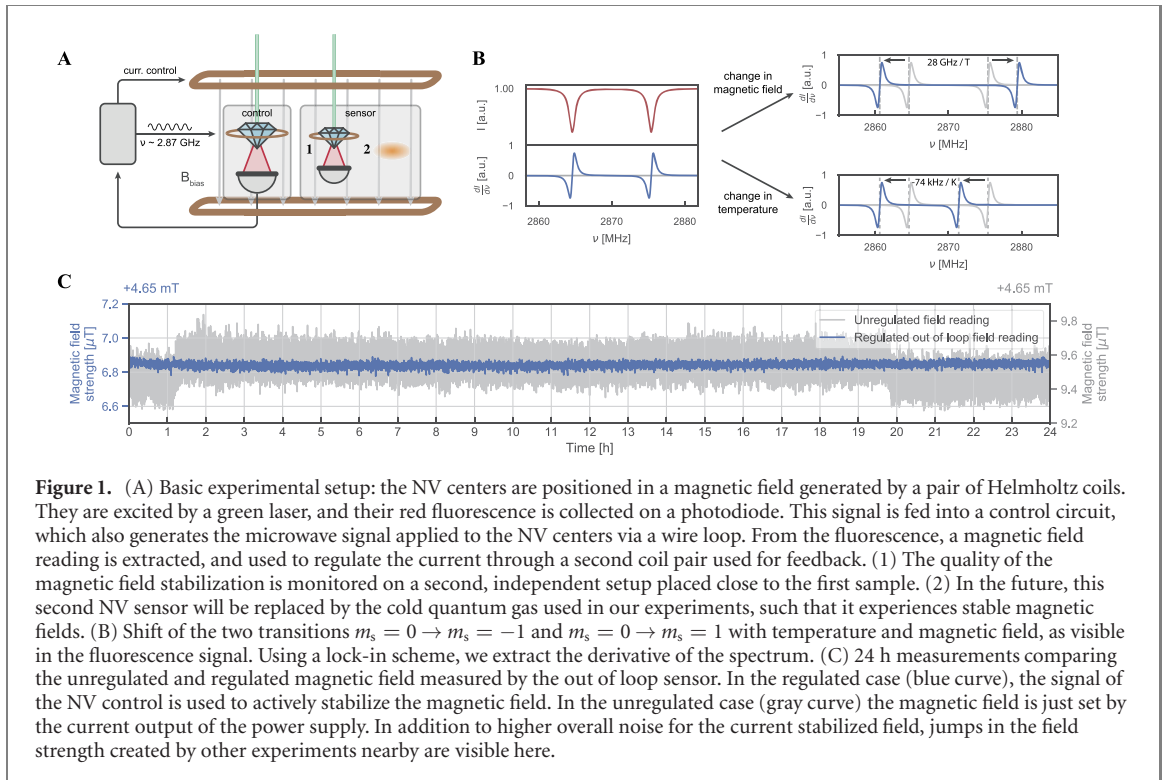
Over the last years, several sensors based on quantum systems with the potential to close this gap have been tested, summarized in table 1. Most prominently, SQUIDs can measure magnetic fields with sensitivities in the $\frac{fT}{\sqrt{Hz}}$ regime [10,11], with atomic gases reaching comparable or slightly better sensitivities at low fields [2,12–14]. However, the large size of these systems as well as the demanding setup required for probing them makes them impractical as a sensor for an active magnetic field stabilization.

Table 1. Comparison of commercially available Hall- and Fluxgate sensors with GMI magnetometers, atomic vapor cells, SQUIDs and NV centers. The extent of the sensor head is used to quantify the optical access to the experiment blocked.

Sensor	Maximum operation field, mT	Sensitivity, $\frac{\mu\text{T}}{\sqrt{\text{Hz}}}$	Blocked optical access, mm	References
Hall sensor	70–200	100–500	1.5–3	[23–25]
Fluxgate	1	5×10^{-3} to 10^{-2}	15–30	[5,26,27]
Giant magnetoimpedance	250×10^{-3} to 1	10^{-3} to 10^{-2}	0.1–20 ^a	[26,27]
Atomic vapor cell	10^{-1}	10^{-7} to 10^{-1}	2–20	[12–14]
SQUID	10^4	10^{-6} to 10^{-2}	10^{-1} to 20 ^b	[26,28–30]
NV center	10^4	9×10^{-4} to 10	10^{-1} to 2	[16–19,31]

^aBased on the orientation of the wire.

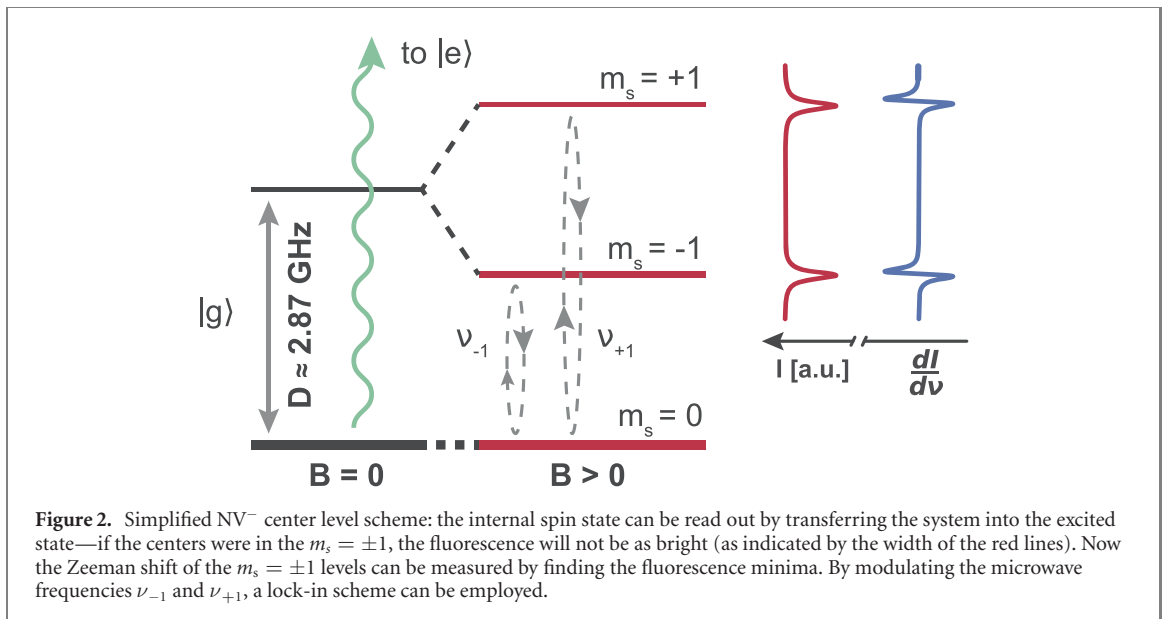
^bRequires cryogenic cooling.



The nitrogen-vacancy (NV) center in diamond provides a versatile and compact magnetic field sensor, covering ranges up to several tens of T with precisions below $1 \frac{\mu\text{T}}{\sqrt{\text{Hz}}}$ [15] for AC magnetic fields. For measurements which approach the DC regime, sensitivities of a few $10 \frac{\mu\text{T}}{\sqrt{\text{Hz}}}$ and below have been reported [16–19]. Due to their small size, they are the perfect candidate for magnetic field stabilizations [20,21] past the range of fluxgate sensors, and allow the study of spin dynamics in regimes previously not experimentally accessible.

In this article, we investigate the direct control of homogeneous magnetic fields based on such NV centers. Our experimental setup is sketched in figure 1: A set of Helmholtz coils produces a homogeneous magnetic field B_0 , which is directly proportional to the applied control current flowing through the coils. This magnetic field is sensed through the fluorescence of a compact NV center magnetometer. In the magnetometer, we measure the optically detected magnetic-resonance (ODMR) features by dressing the ground state spin triplet with a microwave signal of frequency ν . A dip appears in the fluorescence intensity I of the NV centers when ν coincides with one of the Zeeman shifted transition frequencies in this triplet [22]. The derivative $S = \frac{dI}{d\nu}$ is then used as an error signal for the feedback loop. The magnetic field strength has been monitored through a second, independent NV magnetometer over the course of 24 h. We achieved a magnetic field stability of $\sigma_{\text{DC}} = 12 \text{ nT}$ measured over a 1 Hz bandwidth, while shorter measurements over the full control loop bandwidth of 1 kHz yielded magnetic field noise of 126 nT. In the future, replacing this second diamond with the cold atomic clouds of interest will allow us to regulate the magnetic field that the atoms experience.

The remainder of the paper is structured as follows: in section 2 we review the employed detection of the magnetic field and how we separate thermal drifts from magnetic field changes. In section 3 we discuss



the control of the magnetic field and benchmark our system in a detailed fashion. We end with an outlook for more compact solutions beyond this proof-of-concept in section 4.

2. Magnetometry method

For generating the magnetic field signal we employ an ODMR scheme [32] on both our sensors, which we summarize in figure 2: The NV center has a spin $S = 1$ in its ground state. It is optically pumped by a green laser, and consequently emits fluorescence above 637 nm. The unsplit $m_s = \pm 1$ states lie $D \approx 2.87$ GHz [also called the zero field splitting (ZFS)] above the $m_s = 0$ state and hence can be coupled to the ground state through standard microwave techniques (see appendix B). At nonzero magnetic field, the $m_s = \pm 1$ states experience a Zeeman shift of approximately $\pm h\gamma_e \vec{B}_0 \cdot \hat{n}$, where B_0 is the magnetic field strength, $\gamma_e = 28 \text{ GHz T}^{-1}$ is the gyromagnetic ratio of the electron, and \hat{n} is the projection of the magnetic field onto the NV center axis. For the $m_s = \pm 1$ states, a nonradiative decay channel to the ground state leads to a slight decrease in fluorescence, allowing us to extract the transition frequencies ν_{+1} and ν_{-1} from the fluorescence spectrum.

We use a lock-in scheme by frequency modulating the microwave signals applied at modulation frequencies $\omega_{\text{mod}\pm 1}$ of a few tens of kHz to generate the error signals $S_{-1} = \frac{dI}{d\nu}|_{\nu=\nu_{-1}}$ and $S_{+1} = \frac{dI}{d\nu}|_{\nu=\nu_{+1}}$ at both transitions. The information gained by monitoring both transitions can then be used to compensate for temperature fluctuations of the sample (see appendix C), which would shift both resonance frequencies in the same direction, as illustrated in figure 1(B) [33].

The frequency modulated signals necessary for this scheme are generated on a StemLab RedPitaya, and mixed with the two channels of a microwave generator to shift them to the microwave regime to address the two transitions. After that, they are amplified and applied to a wire loop placed around the diamond sample.

The diamond samples used were grown with the HPHT technique, have a natural abundance of ^{13}C and are type 1b diamonds with a natural linewidth of 1.3(1) MHz. The samples are heatsunk to a sapphire window and placed in a homogeneous magnetic field. This is generated by a pair of Helmholtz coils, with a second pair of coils with only few windings (referred to as the feedback coil pair) wound onto, which allows us to regulate the magnetic field over a high bandwidth (see appendix D for further details).

The fluorescence of our sample is collected efficiently by a compound parabolic concentrator [34], and the remaining excitation light is filtered out by a longpass filter (see appendix A for more details). It is measured using an auto-balanced photodetector [35,36], whose output is digitized by the RedPitaya. A modified version of PyRPL [37] is used to process the data on its FPGA, generating an error signal at both transitions to regulate the current through the feedback coil pair.

The position of these transitions depends on the magnetic field, but also on other factors, most importantly temperature. This effect shifts both transitions equally. As the magnetic field shift both lines in opposite directions, we only use the difference signal $\Delta S = S_{+1} - S_{-1}$ as a control signal for the magnetic

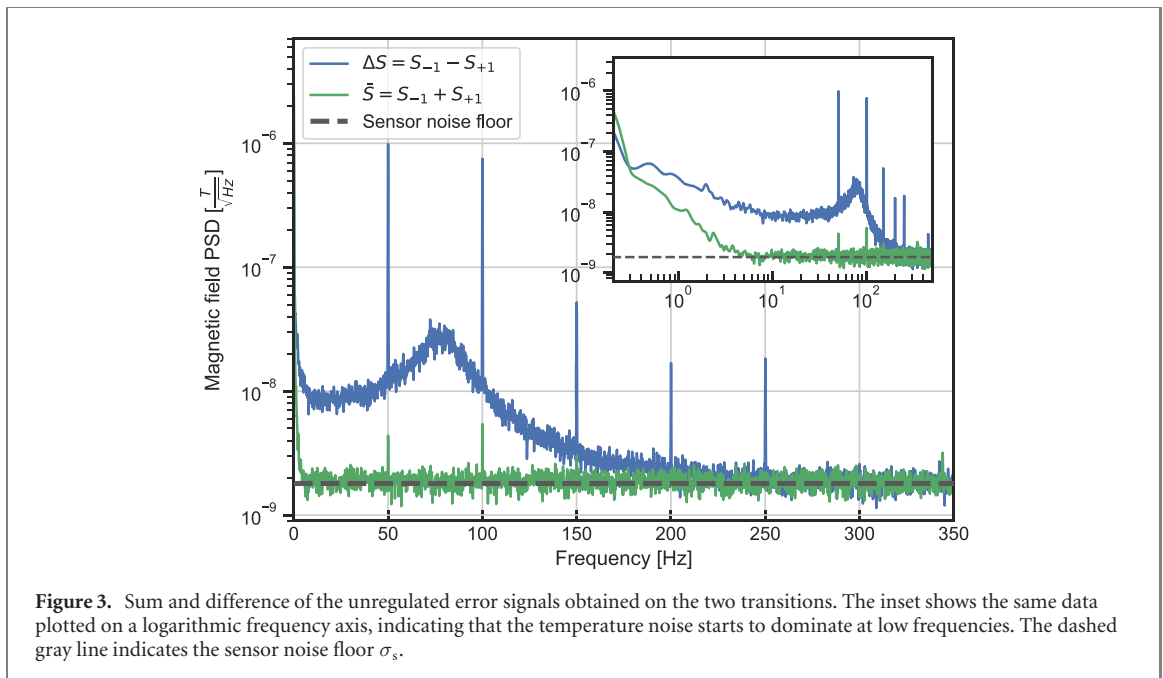


Figure 3. Sum and difference of the unregulated error signals obtained on the two transitions. The inset shows the same data plotted on a logarithmic frequency axis, indicating that the temperature noise starts to dominate at low frequencies. The dashed gray line indicates the sensor noise floor σ_s .

field. The sum of the two error signals $\bar{S} = S_{+1} + S_{-1}$ is directly proportional to the shift of the ZFS D and the off-axis magnetic field.

The spectrum of the unregulated magnetic field obtained this way is shown in figure 3: one can see that the noise at 50 Hz and harmonics, which is magnetic field noise created by the power line, is by a factor 200 smaller in \bar{S} compared to ΔS . From this we conclude that we separate in-axis reading well from off-axis magnetic field and temperature readings, which helps us to stabilize the field in the axis of interest without introducing noise into the system. In the inset we present the same data on a logarithmic frequency scale. Here, one can clearly identify the strong temperature noise, which would eventually become dominant at low frequencies in the individual error signals S_{+1} and S_{-1} .

In total, we collect about 3.4 mW of fluorescence on the signal photodiode for both sensors, leading to a photocurrent of 1.7 mA. This allows us to calculate the shot noise on the photocurrent using $\sigma_{I,SN} = \sqrt{2 \cdot I \cdot e \cdot \Delta f}$ [17]. In this equation, I denotes the photocurrent, e the elementary charge, and Δf is the single-sided measurement bandwidth.

The photocurrent shot noise can be converted to the shot noise limited magnetic field sensitivity $\sigma_{B,SN} = \sigma_{I,SN} \cdot \frac{dB}{dI}$ by multiplying it with the inverse of the lock-in slope. This yields a shot noise limited noise floor of $\sigma_{B,SN} \approx 0.97 \frac{nT}{\sqrt{Hz}}$ for the magnetic field signal, whereas we observe a total noise floor of $\sigma_s \approx 1.8 \frac{nT}{\sqrt{Hz}}$ experimentally. The main additional contributions to this noise floor are not entirely canceled fluorescence intensity noise as well as microwave noise, with a smaller contribution stemming from the shot noise on the reference photocurrent, which was not considered here.

3. Magnetic field stabilization

After the characterization of the magnetic field sensors, we employ them to monitor the magnetic field noise and actively stabilize it. The time trace of these measurements over 24 h was presented in figure 1(C). In figure 4(A) we investigate the associated spectrum at high frequencies. In the open loop configuration (gray curve in the figures), the magnetic field is set to $B_0 = 4.67$ mT by applying a current of ≈ 6.3 A to the pair of Helmholtz coils. The noise spectrum in figure 4(A) reveals that the power supply driving this current introduces magnetic field noise of a few $10 \frac{nT}{\sqrt{Hz}}$ up to a frequency of 200 Hz. Above this cutoff frequency, which is set by the inductance of the coils, the detected magnetic field noise is again limited by the sensor noise.

We use the error signal of the magnetometer to actively stabilize the magnetic field through a feedback system and sense the resulting magnetic field on the independent sensor. The feedback is created by feeding the error signal into a PI controller (which is realized on the RedPitaya's FPGA) that regulates the current through the feedback coil pair. Particularly strong magnetic field noise is present at 50 Hz and harmonics. Because the gain of the PI controller cannot be increased enough to fully cancel this noise, we implement an

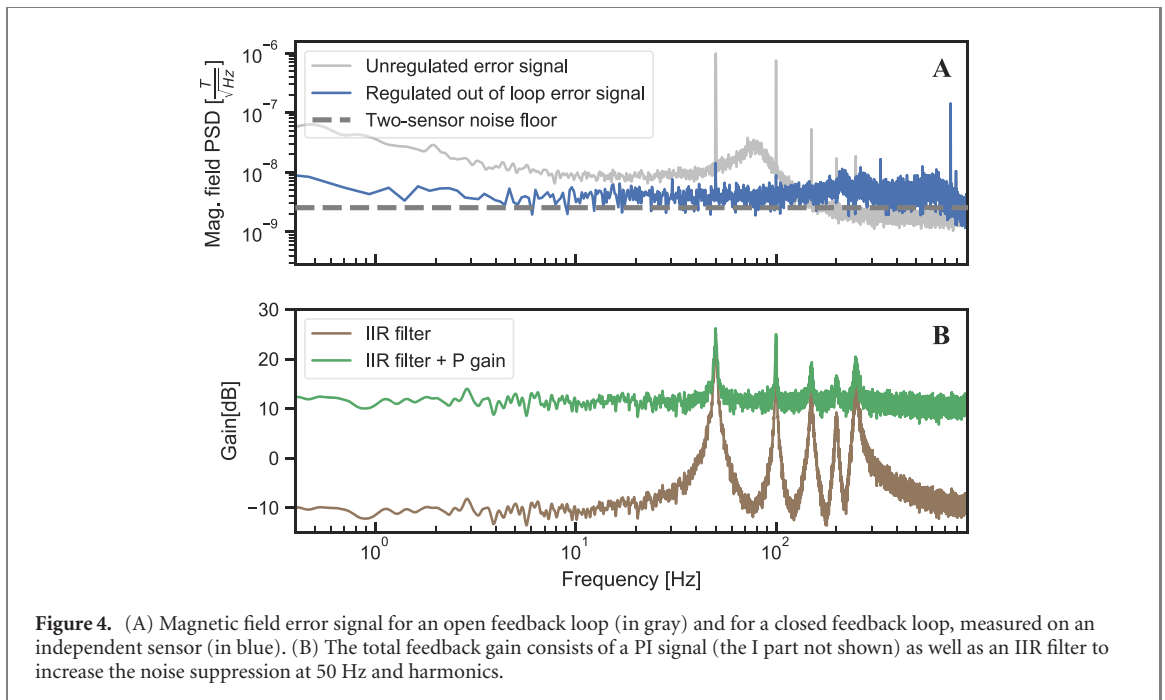


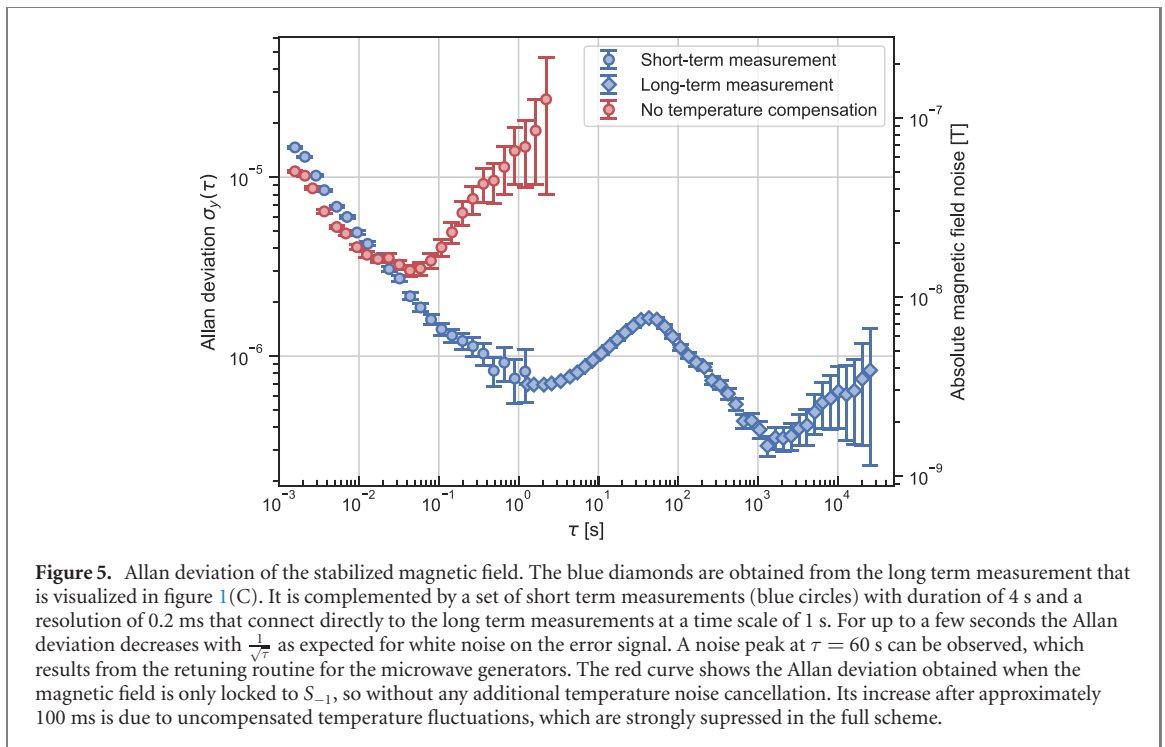
Figure 4. (A) Magnetic field error signal for an open feedback loop (in gray) and for a closed feedback loop, measured on an independent sensor (in blue). (B) The total feedback gain consists of a PI signal (the I part not shown) as well as an IIR filter to increase the noise suppression at 50 Hz and harmonics.

IIR filter to increase the feedback loop gain at these particular frequencies, as shown in figure 4(B). This allows us to fully cancel noise at 50 Hz and harmonics without the feedback loop becoming unstable.

We monitor the resulting magnetic field over 24 h as shown by the blue curve in figure 1(C). Over a bandwidth of 1 Hz we reduce the magnetic field noise of 65 nT down to $\sigma_{DC} = 12$ nT. The DC component of the stability is essential for the repeatability of the quantum simulation experiments, where single runs are typically performed on the timescale of up to a minute and datasets are accumulated over the duration of days [8,9]. The spectrum of a short-term measurement covering the full stabilization bandwidth is shown in figure 4(A): here, we see that up to the control loop's bandwidth of 1 kHz magnetic field noise is reduced down to $\sigma_{AC} = 126$ nT. Within the noise spectrum, peaks that might coincide with a motional resonance are especially problematic. This makes the reduction of the 50 Hz noise by a factor of about 70 a potentially crucial improvement for the fidelity of the quantum simulator.

From these two measurements we calculate the Allan deviation of the magnetic field stability, as shown in figure 5. Additionally, a short-term measurement only stabilized to S_{-1} , so without the temperature noise compensation scheme, is shown, clearly indicating the necessity of this scheme. The short-term measurement with temperature noise compensation drops with the $\tau^{-\frac{1}{2}}$ scaling expected from white frequency noise. The long-term measurements directly connect to the short-term measurements at the integration time of 1 s. For longer integration times we observe a broad peak centered at 60 s. This is due to the dead time during the retuning of the microwave sources once a minute to follow the temperature shifts of the NV center's transitions (see appendix C). We also attribute the slight increase in magnetic field noise at low frequencies in figure 4(A) to these retuning effects. Hence, we see them as the main source of the low frequency noise σ_{DC} . However, this re-tuning interval can be adjusted to the experiment's cycle time, so that a typical atomic physics experiment will not be disturbed by this dead time.

Our work extends previous work on feedback control using NV centers: in the work of [21], feedback from the electron spin of NV centers was used to stabilize the precession of the ^{14}N nuclear spin ensemble associated with them by regulating the detuning of a Ramsey sequence. This allowed to reach longterm stabilities of several hours. Similarly, in [38], a feedback loop tracks resonance pairs of NV centers by supplying feedback to the microwave frequency. This technique offered an increase in dynamic range, and made it robust against temperature fluctuations. In [20] a second NV magnetometer, which was operated close to the sample of an NV-NMR setup, was utilised to stabilize the external magnetic field. The short-term stabilization of up to 5 min showed a magnetic field stability of $\sigma_B \approx 15$ nT over a bandwidth of 12.5 Hz. However, every 5 min, the in-loop sensor signal was zeroed, and the NV-NMR machine was recalibrated to avoid slow drifts of the field. This approach effectively erased information about longterm magnetic field drifts, while the microwave retuning described here only cancels temperature drifts to stay on the resonances, and thus conserves the longterm magnetic field information.



4. Conclusion

In this article we presented a direct magnetic field stabilization based on NV centers, which is compatible with intermediate magnetic fields at the typical scales in cold atom experiments. Our prototype achieved performances which are better than the state-of-the-art for quantum gas experiments at high magnetic fields. The sensitivity of the sensor can be improved further by utilizing better diamond samples, higher collection and excitation efficiencies, increasing the size of the used ensemble [15,16,18], using magnetic flux concentrating materials [19] or microwave cavity readout [39,40]. A fiber based approach to reading out the NV's spin states [41–43] could allow for an even more compact design, which would open the route to applications in confined spaces. For applications with shorter cycle times, or applications which do not allow for a dead time, the signal \bar{S} could also be used to regulate the temperature of the NV sample. With several magnetic field sensors positioned suitably around the cold atom chamber, one might extend the scheme beyond the control of magnetic offset fields and directly control strong magnetic field gradients.

Altogether, this technology could allow for a new class of precision experiments in atomic physics at magnetic field strengths previously not accessible. One route here would be experiments on Feshbach resonances, for instance for the quantum simulation of dipolar droplets [6], where the magnetic field stability would improve by up to two orders of magnitude from $\sim 1 \mu\text{T}$ to the demonstrated $\sigma_{\text{DC}} = 12 \text{ nT}$, or the study of supersolid behaviors at high field values in dipolar quantum gases [44,45]. Another possible application is the study of spin changing collisions between bosonic and fermionic species. For Bose–Bose mixtures, these resonances typically lie in the range of a few hundred μT and have been successfully employed before as a building block of abelian lattice gauge theories [4,46]. To simulate quantum electrodynamics, it will be necessary to employ fermionic atoms, which emulate the role of the fermionic charges. For the scattering of a bosonic and a fermionic species these resonances however do occur at much higher field strengths, e.g. for the case of ^{23}Na and ^{40}K in the vicinity of 30 mT. The demonstrated field stability σ_{DC} results in an energy stability on the order of 100 Hz in this case, which is putting the quantum simulation of quantum electrodynamics into an accessible regime [7].

Finally, the hybrid of ultracold quantum gases and color centers might allow for a new class of experiments at the interface of these two quantum systems. One possible application here would be the cross validation of the previously measured g-factor of the NV center [47–49] with higher precision than before.

Acknowledgments

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Data availability statement

The data that support the findings of this study are openly available at the following URL/DOI: <https://doi.org/10.11588/data/W60KUN> [50].

Appendix A. Optical setup

For both sensors used in this work, the optical setups are identical, with exception of the laser: Either a Laser Quantum Finesse laser or a 520 nm diode laser acquired off ebay supply approximately 900 mW of excitation light. We focus it down onto the diamond samples using an $f = 200$ mm lens, and collect the fluorescence using a compound parabolic concentrator (Edmund Optics, with a 25° collection angle and a 4.3 mm exit diameter). The fluorescence passes a 650 nm longpass filter (Thorlabs FEL0650) to filter out any remaining excitation light before being focused onto the signal photodiode of an autobalanced photodetector (Nirvana Auto-Balanced Photoreceivers Model 2007).

To collect a reference beam for our photodetector, the excitation beam is sent through a $\frac{\lambda}{2}$ waveplate as well as a beam sampler (Thorlabs BSF10-A). The sampled beam passes through a diffuser (Thorlabs DG10-220-A) to remove etalon fringing and similar effects before being detected and used as the reference on the autobalanced photodetector.

Appendix B. Microwave setup and control loop

In order to keep both the microwave setup and the control loop (illustrated on the left in figure 6) compact, much of the functionality was realized on the FPGA of a Red Pitaya STEMLab 125-14 board using a modified version of PyRPL [37]. This allows us to use only few additional components to shift the frequencies into the GHz regime.

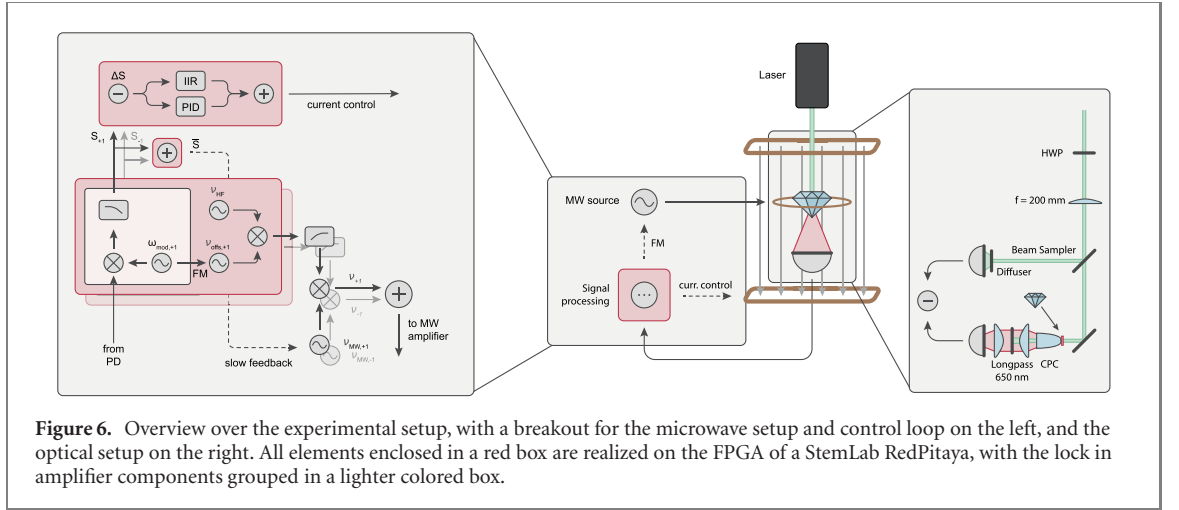
To cancel out the temperature noise in our signal, both transitions $m_s = 0 \rightarrow m_s = -1$ and $m_s = 0 \rightarrow m_s = +1$ are monitored simultaneously. This is possible by frequency modulating the microwave signals driving the transitions at different frequencies $\omega_{\text{mod},-1}$ and $\omega_{\text{mod},+1}$. The two signals are separated by demodulating the fluorescence at said frequencies using a lock-in scheme.

The frequency modulation is generated on a frequency $\nu_{\text{offs},-1}$ as well as $\nu_{\text{offs},+1}$ directly on the RedPitaya's FPGA. Afterward, the signals are mixed with the NV center's hyperfine splitting ν_{HF} to generate sidebands at ≈ 2.2 MHz. These signals are then outputted via the RedPitaya's fast DACs, and high pass filtered afterward (using Mini-Circuits ZFHP-0R50-S+ high pass filters) to remove any low frequency output (due to technical constraints, the current control signal is outputted on the same DAC). This filtered signal is then mixed (Mini-Circuits ZX05-43H-S+) with the microwave frequencies $\nu_{\text{MW},-1}$ and $\nu_{\text{MW},+1}$ (generated by the two channels of a Windfreaktech SynthHD) to shift the frequencies to the GHz regime, creating the frequencies ν_{-1} and ν_{+1} . Finally, the two signals are added together using a power splitter (Mini-Circuits ZX10-2-442-S+) and fed into an amplifier (a Mini Circuits ZHL-16W-43-S+ in-loop or a Mini-Circuits ZHL-15W-422+ out of loop). The amplifier output is applied to the diamond samples via a wire loop glued to the same sapphire glass heatsink as the sample.

Meanwhile, the output of the two lock-in modules S_{-1} and S_{+1} on the RedPitaya is normalized and subtracted (added) together to generate the signals ΔS (\bar{S}). \bar{S} is used in the temperature noise cancellation scheme explained in the next section, while ΔS is fed into both a PI controller as well as into an IIR filter programmed to increase the gain at 50 Hz and harmonics to achieve a stronger noise suppression at those particular frequencies. This allows us to regulate the very strong magnetic field noise at these frequencies to a similar level as the much lower noise at nearby frequencies. Both outputs are added together and outputted on one of the RedPitaya's fast DACs. This control signal is then fed into a home built current controller regulating the current through the feedback coil pair, which stabilizes the magnetic field.

Appendix C. Temperature noise cancellation scheme

Great care has to be taken in order to extract an error signal only containing the magnetic field reading along the axis of interest, without any off-axis components or temperature fluctuation signals.



Conveniently, both off-axis fluctuations as well as temperature fluctuations move the two transitions $m_s = 0 \rightarrow m_s = -1$ and $m_s = 0 \rightarrow m_s = 1$ in unison, while the on axis fluctuations move them in opposite directions, as illustrated in figure 1(B). This allows to extract the on-axis magnetic field reading by subtracting the normalized error signals generated on both transitions independently. To normalize these signals, magnetic field fluctuations at a frequency with a low noise floor were generated on the feedback coil pair, and their effect on the sum signal \bar{S} was minimized by scaling the two error signals accordingly.

One issue that can occur in this configuration is that the temperature drifts far enough to shift the transitions away from the two microwave frequencies ν_{-1} and ν_{+1} —luckily, this can be circumvented by observing the sum signal \bar{S} and retuning the absolute microwave frequencies (while keeping their relative distance, which sets the magnetic field strength), keeping \bar{S} near zero.

A crude feedback loop—consisting of several measurements of \bar{S} followed by a retuning of the microwave frequencies in order to regulate the sum signal—was implemented for the measurements presented in this work. Even though it generates some noise for every retuning (as can be seen on the peak at 60 s in figure 5), it is essential to ensure the long-term functionality of the feedback loop, and to prevent it from falling out of lock.

For applications where the dead time due to the microwave re-tuning is not acceptable, a temperature stabilization of the sample would be possible. Here however, care needs to be taken such that the current of the heating/cooling element does not create magnetic field noise or magnetic field gradients near the sample. Alternatively, modulating the laser intensity might be a viable option, keeping the sensor head compact at the cost of a slightly fluctuating lock-in slope $\frac{dI}{dB}$.

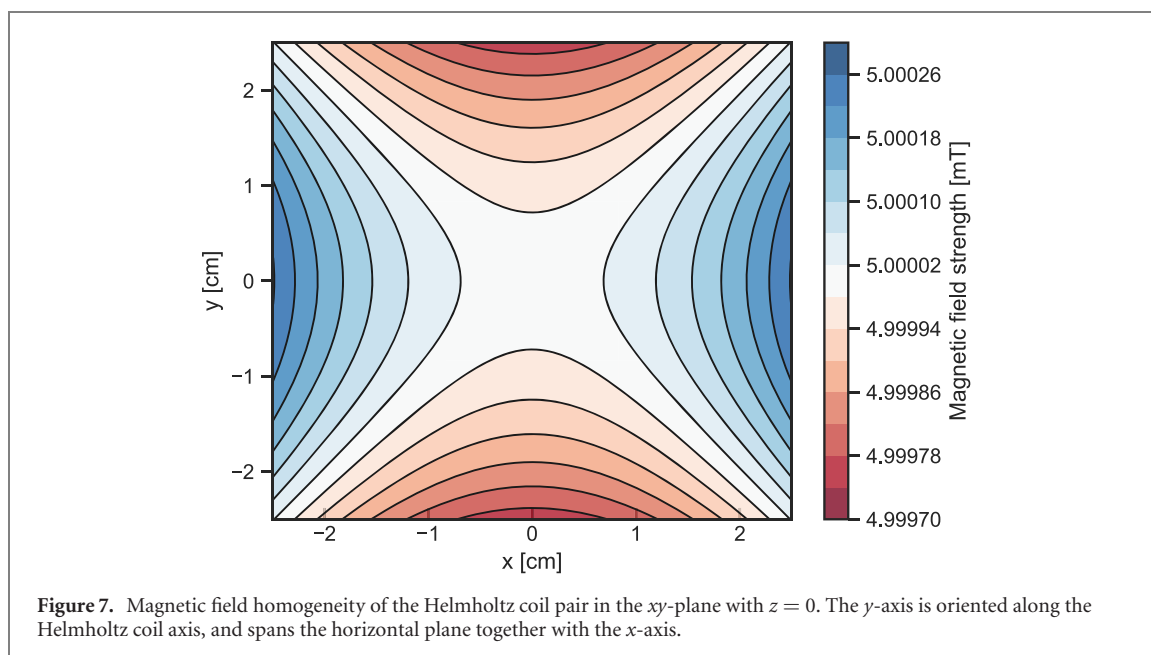
Appendix D. Magnetic field generation

We employ a pair of coils of side lengths 33.5 cm \times 28.5 cm with 80 windings each and separated by 15.5 cm to generate our magnetic bias field. The feedback coil pair has 16 windings wrapped around the first coil pair, and allows us to regulate the magnetic field over a larger bandwidth due to its low inductance. While the bias coil pair is driven by a Delta Elektronika SM 18-50, this coil pair is regulated using a home built current driver.

One requirement for the magnetic field in our application is to be homogeneous over both sensors to ensure that the signals read out are identical. For this, the coil geometry used was simulated using Radia [51], yielding inhomogeneities below 50 ppm for 1 cm total misalignment of the two sensors with respect to the coil pair's center (figure 7). As the AC magnetic field fluctuations we want to stabilize are on the order of $1 \frac{\mu\text{T}}{\sqrt{\text{Hz}}}$ and below, this is currently not limiting. However, magnetic field gradients leading to different sensor readings can also be generated by external sources, like 50 Hz power line noise, making it desirable to keep both sensors as close together as possible.

In a similar manner, the magnetic field inside the diamond samples was simulated. As inhomogeneities here lead to a broadening of the resonances for large magnetic fields, the sensitivity S can be calculated using the magnetic field spread ΔB and the system's transition linewidth at zero magnetic field Δf :

$$S \propto \sqrt{\frac{\Delta f}{\sqrt{\Delta f^2 + \Delta B^2 \cdot \gamma_e^2}}} \quad (\text{D1})$$



Based on this scaling, we expect the sensitivity of our sensor to decrease by 50 % past a field value of 2.2 T.

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